

A Rate-Compatible Sphere-Packing Analysis of Feedback Coding with Limited Retransmissions

Adam Williamson, Tsung-Yi Chen, and Rick Wesel

UCLA Communication Systems Laboratory
arXiv: 1202.1458

July 6, 2012

Variable-Length Feedback with Termination

Variable-Length Feedback with Termination

- **Variable-length feedback with termination (VLFT)** codes [Polyanskiy et al. 2011]:
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.

Variable-Length Feedback with Termination

- **Variable-length feedback with termination (VLFT)** codes [Polyanskiy et al. 2011]:
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may terminate after each symbol.

Variable-Length Feedback with Termination

- **Variable-length feedback with termination (VLFT)** codes [Polyanskiy et al. 2011]:
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may terminate after each symbol.
 - ⇒ (Rate-compatible) random coding.

Variable-Length Feedback with Termination

- **Variable-length feedback with termination (VLFT)** codes [Polyanskiy et al. 2011]:
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may terminate after each symbol.
 - ⇒ (Rate-compatible) random coding.
 - ⇒ General results with numerical examples for BSC and BEC.

This Talk

- **This talk:** Still the basic VLFT framework.
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.

This Talk

- **This talk:** Still the basic VLFT framework.
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may only terminate at the end of a “packet”.

This Talk

- **This talk:** Still the basic VLFT framework.
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may only terminate at the end of a “packet”.
 - ⇒ Incremental packet lengths will be optimized.

This Talk

- **This talk:** Still the basic VLFT framework.
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may only terminate at the end of a “packet”.
 - ⇒ Incremental packet lengths will be optimized.
 - ⇒ Rate-compatible sphere-packing (RCSP) [Chen et al. 2011].

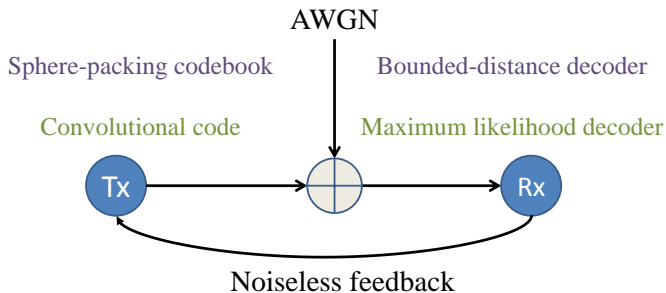
This Talk

- **This talk:** Still the basic VLFT framework.
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may only terminate at the end of a “packet”.
 - ⇒ Incremental packet lengths will be optimized.
 - ⇒ Rate-compatible sphere-packing (RCSP) [Chen et al. 2011].
 - ⇒ Rate-compatible tail-biting convolutional code.

This Talk

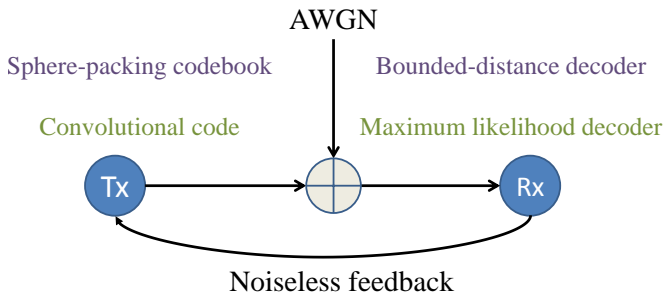
- **This talk:** Still the basic VLFT framework.
 - ⇒ Transmitter sees all channel outputs and tells the receiver when to terminate through a separate control channel.
 - ⇒ Transmission may only terminate at the end of a “packet”.
 - ⇒ Incremental packet lengths will be optimized.
 - ⇒ Rate-compatible sphere-packing (RCSP) [Chen et al. 2011].
 - ⇒ Rate-compatible tail-biting convolutional code.
 - ⇒ Focused on the AWGN channel.

Incremental Redundancy Scheme Overview



- Forward channel is **AWGN** with known SNR, η .

Incremental Redundancy Scheme Overview



- Forward channel is **AWGN** with known SNR, η .
- The receiver attempts to decode after each incremental transmission, based on all received symbols.

Transmission Scheme Details (1st transmission)

- $k = \log_2 M =$ information bits.



- 1st transmission:
 - Send I_1 , decode with $N_1 = I_1$.
 - $R_1 = k/N_1 =$ initial code rate.

Transmission Scheme Details (2nd transmission)

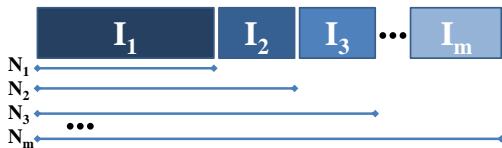
- $k = \log_2 M =$ information bits.



- **2nd transmission:**
 - Send I_2 , decode with $N_2 = I_1 + I_2$.
 - $R_2 = k/N_2 =$ code rate.

Transmission Scheme Details (*i*th transmission)

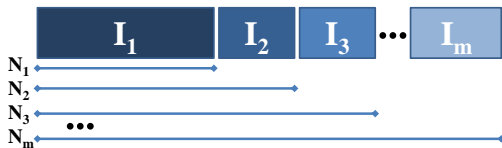
- $k = \log_2 M =$ information bits.



- *i*th transmission: ($i = 2, \dots, m$)
 - Send I_i , decode with $N_i = N_{i-1} + I_i$.
 - $I_i =$ incremental step size, $N_i =$ block length at *i*th transmission.
 - $R_i = k/N_i =$ code rate at *i*th transmission

Transmission Scheme Details (*i*th transmission)

- $k = \log_2 M =$ information bits.

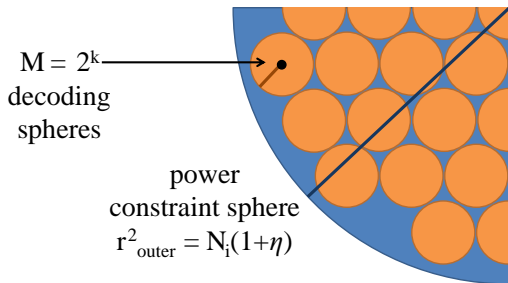


- *i*th transmission: ($i = 2, \dots, m$)
 - Send I_i , decode with $N_i = N_{i-1} + I_i$.
 - $I_i =$ incremental step size, $N_i =$ block length at *i*th transmission.
 - $R_i = k/N_i =$ code rate at *i*th transmission
- $m =$ maximum number of transmissions (before repetition).

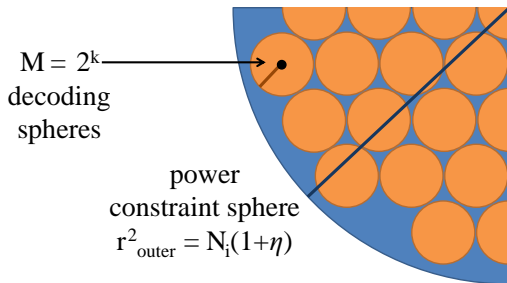
Start Over if Failure after m Transmissions

- If decoding is unsuccessful after m transmissions, start over by sending I_1 bits, then I_2 bits, etc. (similar to ARQ).
- This is a practical limitation.
- Simplifies analysis.

Decoding Error Probability for Sphere-Packing



Decoding Error Probability for Sphere-Packing



- $P[\text{error with block length } N_i]$

$$= P(\zeta_i) = P\left(\sum_{\ell=1}^{N_i} z_{\ell}^2 > r_i^2\right) = 1 - F_{\chi_{N_i}^2}(r_i^2),$$

- $r_i^2 = \frac{N_i(1+\eta)}{2^{2k/N_i}}$ is the sphere-packing radius (squared),
- $z_{\ell} \sim \mathcal{N}(0, 1)$ are the noise samples.

Sphere-Packing: Myth or Reality?

- An ideal sphere-packing codebook is mythical.
 - ⇒ Upper bound on packing density ϕ in n dimensions:
$$\phi \leq (n/e) 2^{-n/2}.$$

Sphere-Packing: Myth or Reality?

- An ideal sphere-packing codebook is mythical.
 - ⇒ Upper bound on packing density ϕ in n dimensions:
$$\phi \leq (n/e) 2^{-n/2}.$$
- ... but we will see that a convolutional code can achieve sphere-packing performance.

Marginal vs. Joint Decoding Error Probability

- $P[\text{error with block length } N_i] = P(\zeta_i)$ **(marginal)**

$$= P\left(\sum_{\ell=1}^{N_i} z_{\ell}^2 > r_i^2\right) = 1 - F_{\chi_{N_i}^2}(r_i^2),$$

Marginal vs. Joint Decoding Error Probability

- $P[\text{error with block length } N_i] = P(\zeta_i)$ **(marginal)**

$$= P\left(\sum_{\ell=1}^{N_i} z_{\ell}^2 > r_i^2\right) = 1 - F_{\chi_{N_i}^2}(r_i^2),$$

- $P[\text{error after } j \text{ transmissions}] = P(\zeta_1, \zeta_2, \dots, \zeta_j)$ **(joint)**

$$\begin{aligned} &= P\left(\sum_{\ell=1}^{N_1} z_{\ell}^2 > r_1^2, \sum_{\ell=1}^{N_2} z_{\ell}^2 > r_2^2, \dots, \sum_{\ell=1}^{N_j} z_{\ell}^2 > r_j^2\right) \\ &= \int_{r_1^2}^{\infty} \int_{r_2^2 - t_1}^{\infty} \dots \int_{r_{j-1}^2 - \sum_{i=1}^{j-2} t_i}^{\infty} f_{\chi_{N_1}^2}(t_1) \dots f_{\chi_{N_{j-1}}^2}(t_{j-1}) \times \\ &\quad \left(1 - F_{\chi_{N_j}^2}\left(r_j^2 - \sum_{i=1}^{j-1} t_i\right)\right) dt_{j-1} \dots dt_1. \end{aligned}$$

Latency and Throughput (for $m = 1$, the ARQ Case)

- $\lambda = \text{latency} =$ expected number of forward channel uses.

Latency and Throughput (for $m = 1$, the ARQ Case)

- $\lambda = \text{latency}$ = expected number of forward channel uses.

$$\begin{aligned}\lambda &= I_1 (1 + P(\zeta_1) + P(\zeta_1)^2 + P(\zeta_1)^3 + \dots) \\ &= \frac{I_1}{1 - P(\zeta_1)} \\ &= \frac{I_1}{F_{\chi_{M_1}^2}(r_1^2)}\end{aligned}$$

Latency and Throughput (for $m = 1$, the ARQ Case)

- $\lambda = \text{latency}$ = expected number of forward channel uses.

$$\begin{aligned}\lambda &= I_1 (1 + P(\zeta_1) + P(\zeta_1)^2 + P(\zeta_1)^3 + \dots) \\ &= \frac{I_1}{1 - P(\zeta_1)} \\ &= \frac{I_1}{F_{\chi_{N_1}^2}(r_1^2)}\end{aligned}$$

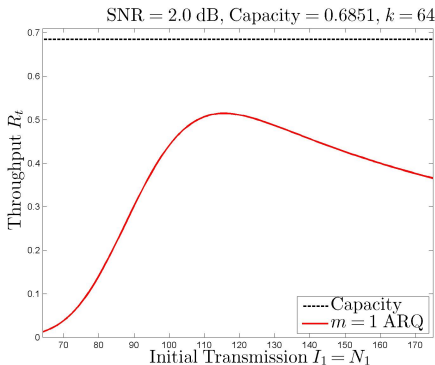
- $R_t = \text{throughput} = k/\lambda$.
- Select I_1 to maximize R_t .

Latency and Throughput (for $m = 1$, the ARQ Case)

- $\lambda = \text{latency}$ = expected number of forward channel uses.

$$\begin{aligned}\lambda &= I_1 (1 + P(\zeta_1) + P(\zeta_1)^2 + P(\zeta_1)^3 + \dots) \\ &= \frac{I_1}{1 - P(\zeta_1)} \\ &= \frac{I_1}{F_{\chi_{N_1}^2}(r_1^2)}\end{aligned}$$

- $R_t = \text{throughput} = k/\lambda$.
- Select I_1 to maximize R_t .



What About $m > 1$?

- Latency

$$\lambda = \frac{I_1 + \sum_{i=2}^m I_i P \left(\bigcap_{j=1}^{i-1} \zeta_j \right)}{1 - P \left(\bigcap_{j=1}^m \zeta_j \right)}$$

What About $m > 1$?

- **Latency**

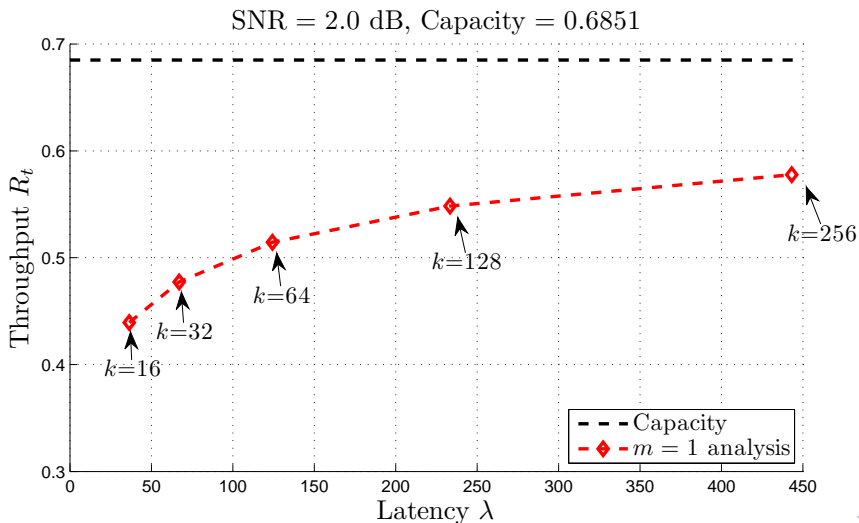
$$\lambda = \frac{I_1 + \sum_{i=2}^m I_i P \left(\bigcap_{j=1}^{i-1} \zeta_j \right)}{1 - P \left(\bigcap_{j=1}^m \zeta_j \right)}$$

- **Throughput**

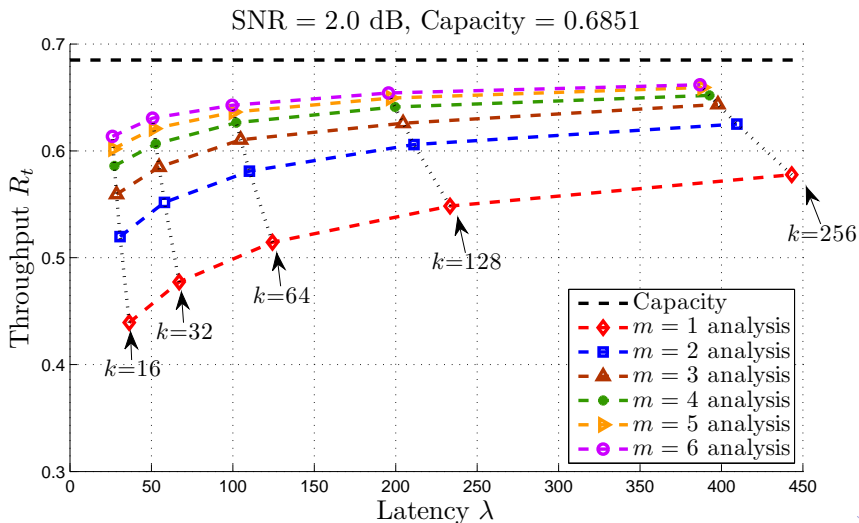
$$R_t = k/\lambda$$

- Select $\{I_1, I_2, \dots, I_m\}$ to maximize R_t .

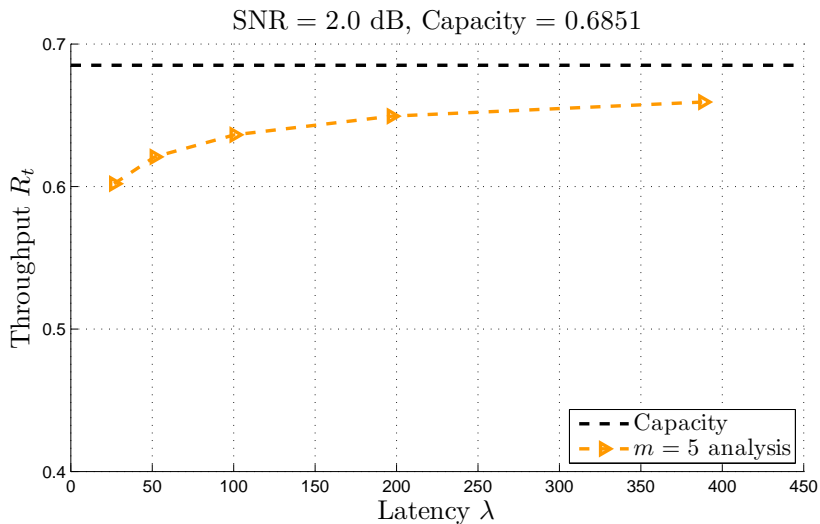
RCSP: Latency vs. Throughput for $m = 1$ (ARQ) Using Optimal Step Size I_1



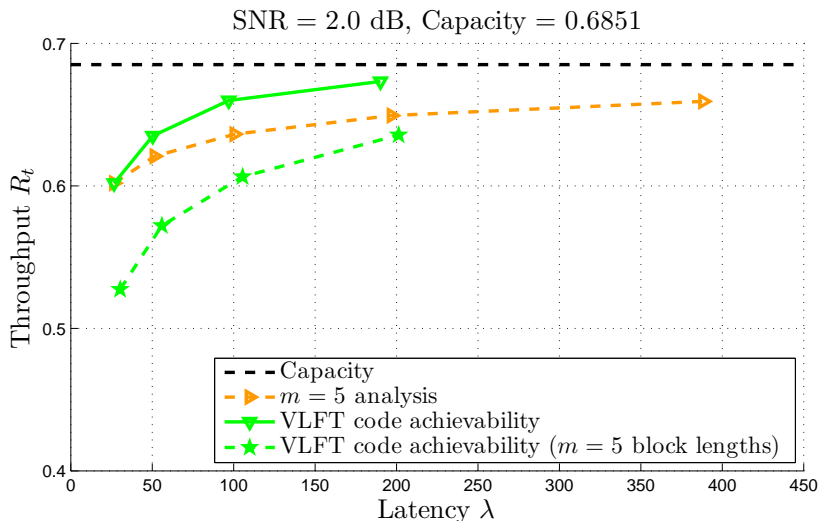
RCSP: Latency vs. Throughput for $m = 1$ to $m = 6$, Using Optimal Step Sizes I_i



RCSP: Latency vs. Throughput for $m = 5$, Using Optimal Step Sizes I_i



Comparison with [Polyanskiy et al. 2011]



Convolutional Code Simulations for $m = 5$

- Mother codes are rate $1/3$, 64-state and 1024-state convolutional codes from [Lin and Costello 2004].

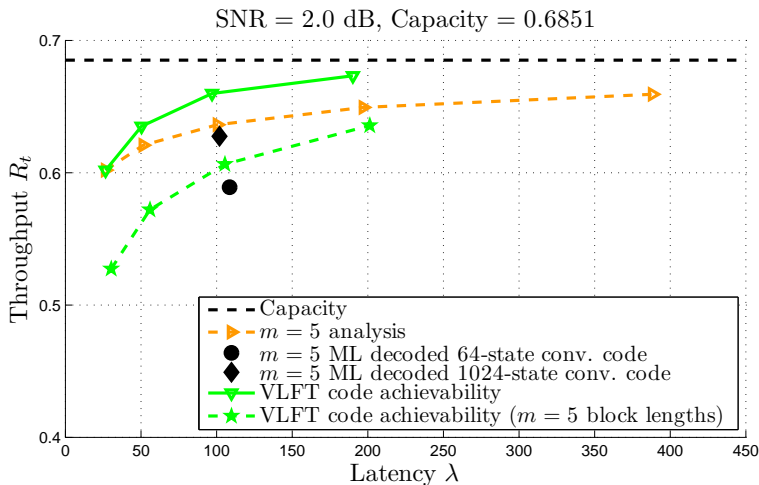
Convolutional Code Simulations for $m = 5$

- Mother codes are rate $1/3$, 64-state and 1024-state convolutional codes from [Lin and Costello 2004].
- Use transmission lengths $\{I_1^m\}$ identified in RCSP optimization for $m = 5$.
- High-rate codes obtained by pseudo-random puncturing of mother codes.

Convolutional Code Simulations for $m = 5$

- Mother codes are rate $1/3$, 64-state and 1024-state convolutional codes from [Lin and Costello 2004].
- Use transmission lengths $\{I_1^m\}$ identified in RCSP optimization for $m = 5$.
- High-rate codes obtained by pseudo-random puncturing of mother codes.
- **Maximum likelihood (ML)** decoding.
 - ML decoding regions completely fill the power constraint sphere.
- Tail-biting implementations used for throughput efficiency.

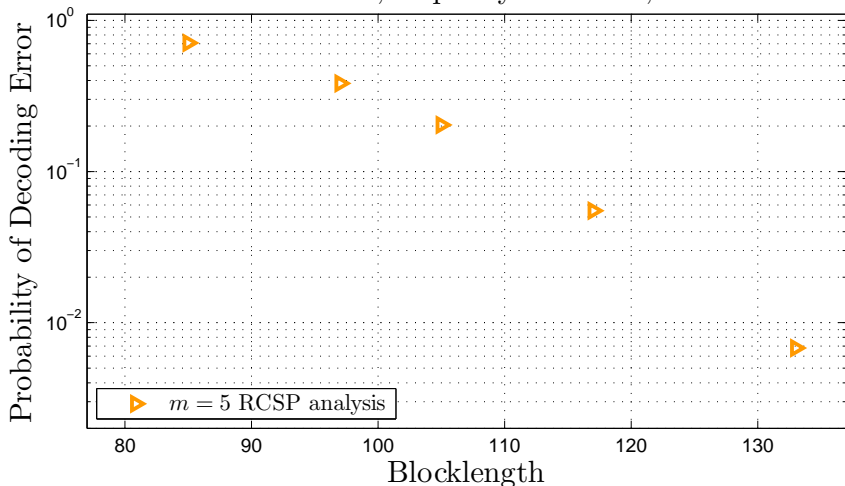
Convolutional Code Achievability, $m = 5$



- 90% of AWGN capacity in ~ 100 symbols.

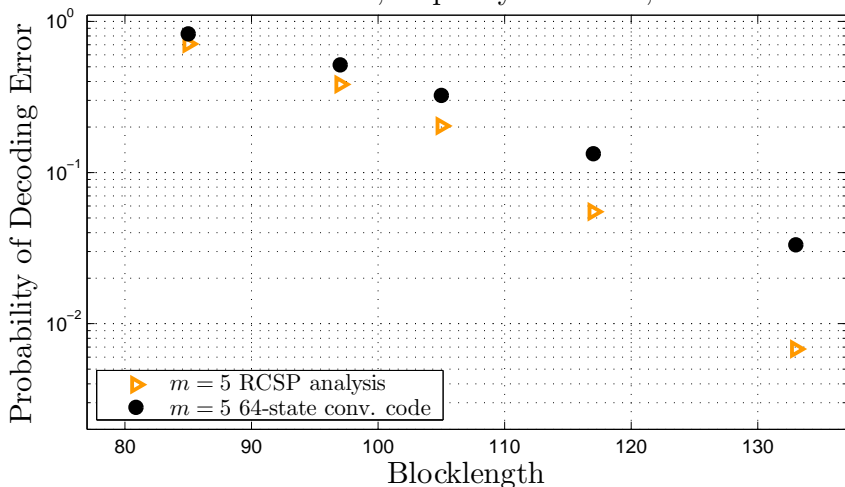
Decoding Error Trajectory

SNR = 2.0 dB, Capacity = 0.6851, $k = 64$



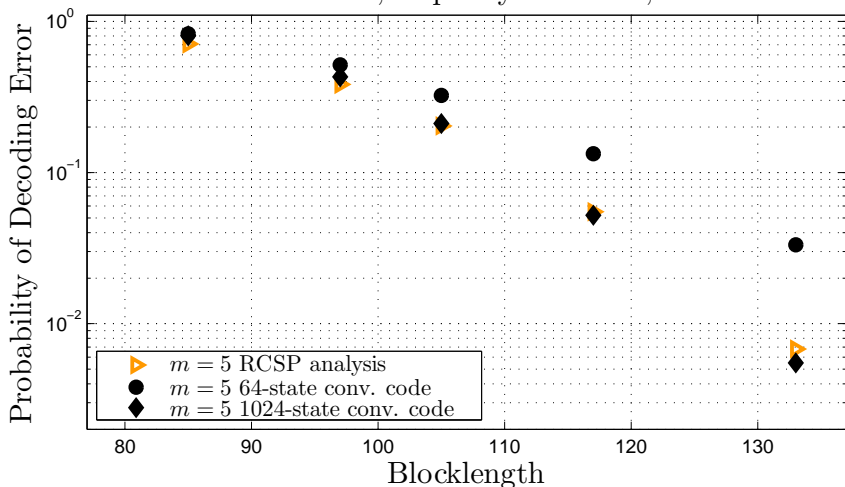
Decoding Error Trajectory

SNR = 2.0 dB, Capacity = 0.6851, $k = 64$



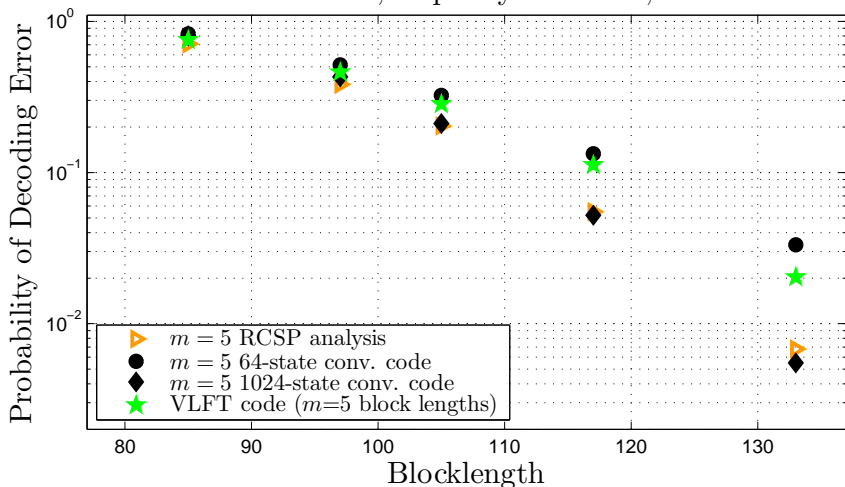
Decoding Error Trajectory

SNR = 2.0 dB, Capacity = 0.6851, $k = 64$



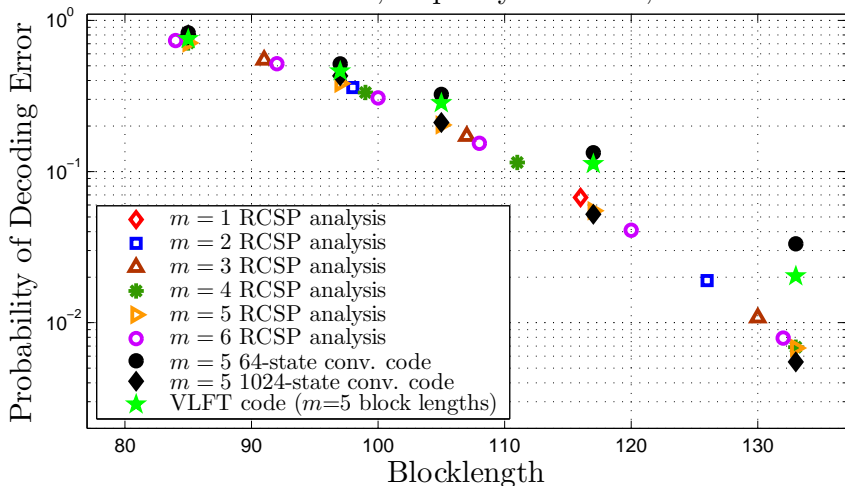
Decoding Error Trajectory

SNR = 2.0 dB, Capacity = 0.6851, $k = 64$



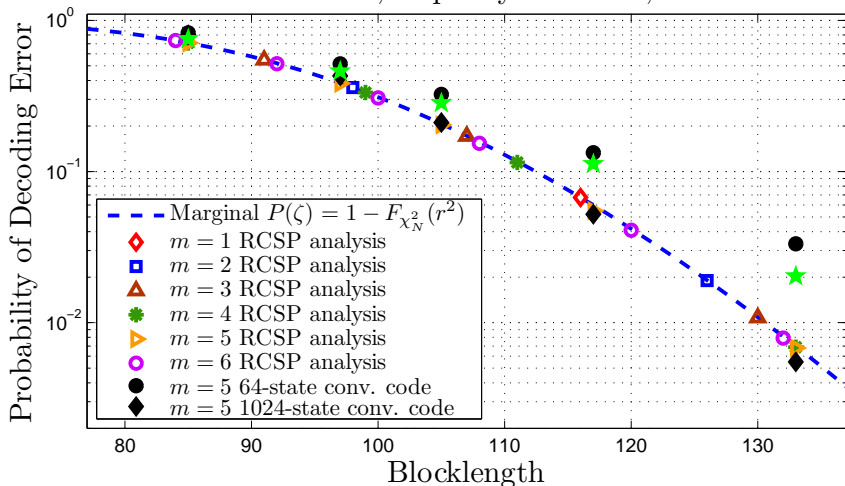
Decoding Error Trajectory

SNR = 2.0 dB, Capacity = 0.6851, $k = 64$



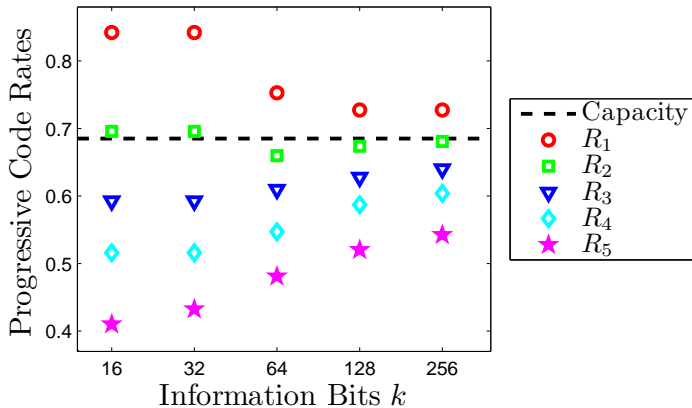
Marginal Chi-Square: A Design Objective

SNR = 2.0 dB, Capacity = 0.6851, $k = 64$



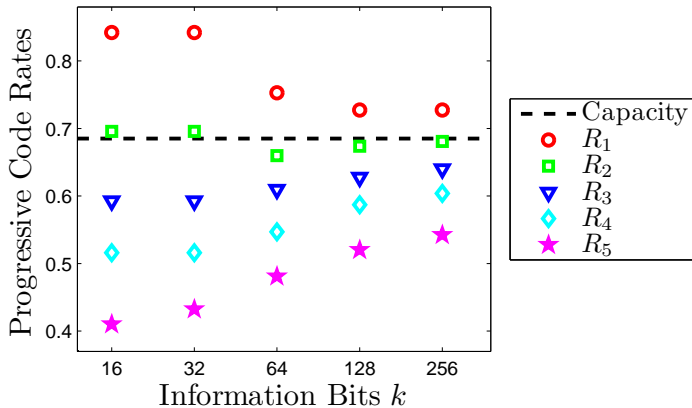
Optimal Rates

SNR = 2.0 dB, Capacity = 0.6851, $m = 5$



Optimal Rates

SNR = 2.0 dB, Capacity = 0.6851, $m = 5$



- $R_1 > C$

Concluding Thoughts

- Feedback improves achievable rate for finite block lengths.

Concluding Thoughts

- Feedback improves achievable rate for finite block lengths.
 - Feedback after every bit is best.

Concluding Thoughts

- Feedback improves achievable rate for finite block lengths.
 - Feedback after every bit is best.
 - When transmissions must be grouped, pick the sizes wisely.

Concluding Thoughts

- Feedback improves achievable rate for finite block lengths.
 - Feedback after every bit is best.
 - When transmissions must be grouped, pick the sizes wisely.
- Find good codes by matching RCSP error trajectories.

Concluding Thoughts

- Feedback improves achievable rate for finite block lengths.
 - Feedback after every bit is best.
 - When transmissions must be grouped, pick the sizes wisely.
- Find good codes by matching RCSP error trajectories.
- Questions?